

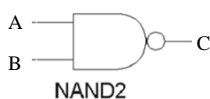
VHDL et la synthèse de circuits séquentiels

Règles de conception

- Adopter une démarche qualité (dénomination, hiérarchisation, testbench...)
- Attention aux assignations incomplètes (mémoire implicite)
- Attention à l'utilisation des variables
- Penser implantation (orienter le compilateur) : il faut connaître l'architecture du composant
- Trouver le bon compromis entre ressources et vitesses en choisissant le bon degré de parallélisation (pas compact mais rapide) ou de sérialisation (compact mais lent)

Qu'est-ce que la Synthèse?

- Générer du hardware depuis HDL
- Eg. $C \leftarrow A \text{ nand } B$ donne



- Pas si simple!
- Nécessite un processus complexe afin d'optimiser un arrangement entre le temps et l'espace.

Danger!

- | | |
|--|--|
| <ul style="list-style-type: none"> • Time expressions – after • Assert, Text I/O • Configuration declarations • Dynamic Loops • Initial values • Sensitivity lists | <p>} Simulation seulement</p> <p>} Modifie les résultats de synthèse</p> |
|--|--|

Règles de conception

- Systèmes synchrones (simu fonctionnelle simplifiée et puissante, analyse temporelle statique possible.) :
- 1 SEULE horloge (broche globale du composant)
- Si plusieurs domaines d'horloges, il faut des FIFOs tampons
- resynchroniser tous les signaux asynchrones pour éviter la métastabilité entre 1 et 0

Eléments séquentiels de base

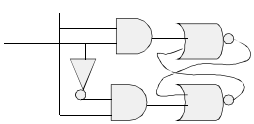
- **Latches**
 - Moins complexe que FFs
 - Perturbe les synchronisations et l'analyse temporelle
 - Des latches sont parfois générés sans le savoir!!
- **Flip-flops**
 - Edge sensitive, front montant ou descendant
 - Asynchronous et synchronous set and reset

Description with Latches

- Data input: D; enable: LE; output: Q.

```

signal D, LE, Q: bit ;
...
bl : process (D, LE)
begin
  if (LE = '1') then
    Q <= D;
  end if ;
end process ;
    
```



Observe that this syntax example involves storing the signal in a memory

Description using Edge Triggered Flip Flops

- Flip Flop is based on a clock signal.
- Rising/Falling edge triggered.

```

signal D, Q, clk : bit ;
....
process (clk)
begin
  if (clk'event and clk='1') then
    Q <= D;
  end if ;
end process ;
    
```

Rising edge

Modeling Latches

- when: Q <= D when En='1' else Q; ... Q <= D when En='1' else unaffected; (VHDL 93)
- if controlled by Enable

```

library IEEE;
use IEEE.Std_Logic_1164.all;
entity Latch is
  port (D, En : in Std_Logic;
        Q : out Std_Logic);
end Latch;

architecture RTL of Latch is
begin
  L1: process(D, En)
  begin
    if En='1' then Q <= D;
    end if;
  end process;
end RTL;
    
```

Descriptions of Synchronous Set/Reset Flip Flops

- Data and reset is NOT on sensitivity list.

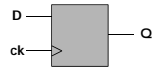
```

signal D, Q, clk, reset : bit ;
...
process (clk)
begin
  if (clk'event and clk = '1') then
    if reset = '1' then
      D <= '0';
    else
      Q <= D ;
    end if ;
  end if ;
end process ;
    
```

D is symbol of next state, not current state

LOGI QUE SEQUENTIELLE / bascules

bascule D « edge-triggered »



1^{ère} possibilité de description:

- utilisation de l'instruction WAIT
- le PROCESS n'est exécuté que lors du passage de 0 à 1 du signal d'horloge
- le PROCESS ne comporte pas de liste de sensibilité
- il est nécessaire de passer par un signal interne (« memo » dans l'exemple)

```

ENTITY DFF IS
  PORT (D,ck : IN BIT;
        Q : OUT BIT);
END DFF ;
ARCHITECTURE arch1 OF DFF IS
  SIGNAL memo : BIT;
  BEGIN
    PROCESS
      BEGIN
        WAIT UNTIL ck='1';
        memo <= D;
      END PROCESS;
      Q <= memo;
    END arch1;
    
```

Asynchronous Set/Reset Flip Flops

- Reset is ON sensitivity list.

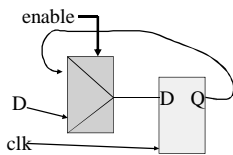
```

signal D, Q, clk, reset : bit ;
...
process (clk, reset)
begin
  if (reset = '1') then
    Q <= '0' ;
  elsif (clk'event and clk = '1') then
    Q <= D ;
  end if ;
end process ;
    
```

Clock Enable Flip Flops

```

signal D, Q, enable, clk : bit ;
...
process (clk)
begin
if (clk'event and clk='1') then
if (enable='1') then
Q <= D ;
end if ;
end if ;
end process ;
    
```



Descriptions of Tristate Buffers

- 3 output states: 0, 1, Z (high impedance)

```

entity tristate is
port ( D, en: in std_logic ;
      Q : out std_logic);
end tristate ;

architecture EG of tristate is
begin
Q <= D when en = '1' else 'Z' ;
end EG ;

p1 : process (en, D)
begin
if (en='1') then
Q <= D ;
else
Q <= 'Z' ;
end if ;
end process ;
    
```

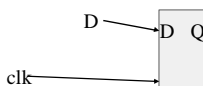
Observe that because I used symbol Z the system inferred tri-state circuit by itself

Flip Flops inferred by Wait Until

```

signal D, Q, clk : bit ;
...
process
begin
wait until clk'event and clk='1' ;
Q <= D ;
end process ;
    
```

- No sensitivity lists
- No asynchronous reset



Description of wired circuit using Tristate Buffers

- Simultaneous assignment to 1 signal
- Does not verify exclusivity

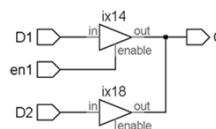
We do not specify if this is wired OR or wired AND or what

```

library IEEE;
use IEEE.std_logic_1164.all;

entity tristate is
port ( D1, D2, en1, en2 : in std_logic ;
      Q : out std_logic);
end tristate ;

architecture EG of tristate is
begin
Q <= D1 when en1 = '1' else 'Z' ;
Q <= D2 when en2 = '1' else 'Z' ;
end EG ;
    
```



Why so many FF syntax?

- Highlight the importance of writing good HDL code
- Coding style affects synthesized structure directly
- Internal working of synthesizers

Busses with tri-state buffers

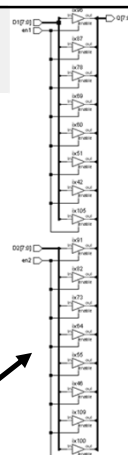
- Use arrays in the declarations

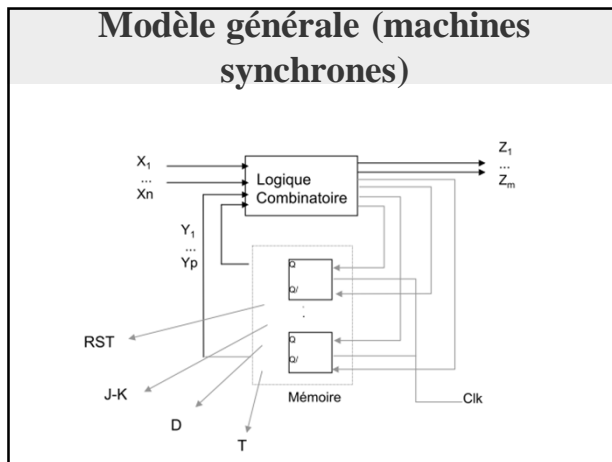
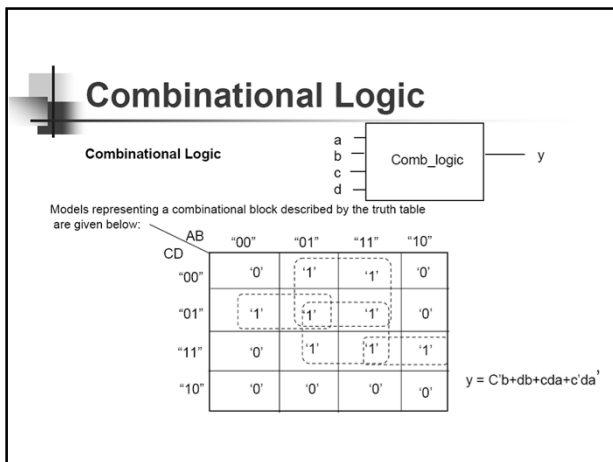
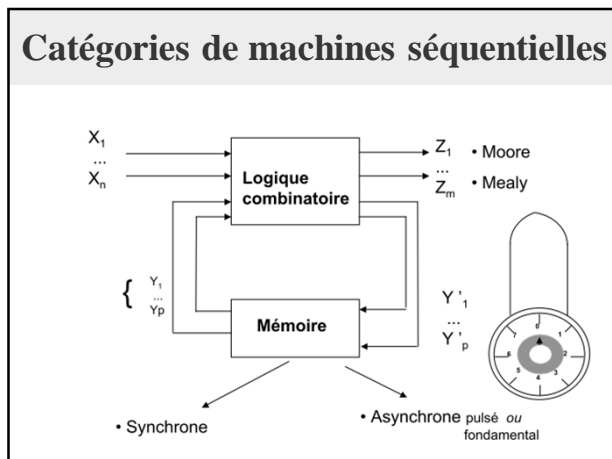
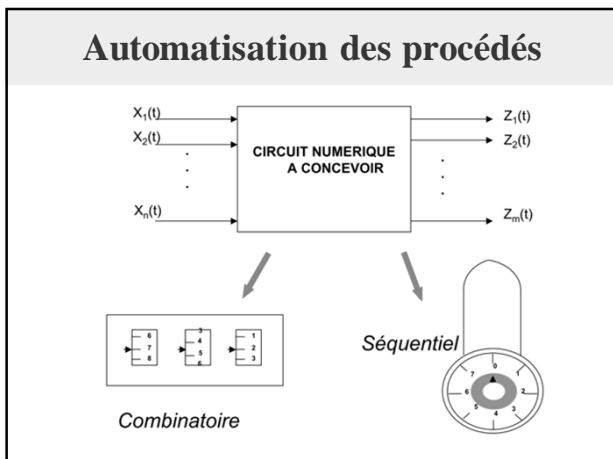
```

entity tristate is
port ( D1, D2: in std_logic_vector (7 downto 0);
      en1, en2 : in std_logic;
      Q : out std_logic_vector (7 downto 0) );
end tristate ;

architecture EG of tristate is
begin
Q <= D1 when en1 = '1' else "ZZZZZZZZ" ;
Q <= D2 when en2 = '1' else "ZZZZZZZZ" ;
end EG ;
    
```

This description and circuit describes selector circuit realized with tri-state buffers on a bus – this is an important circuit

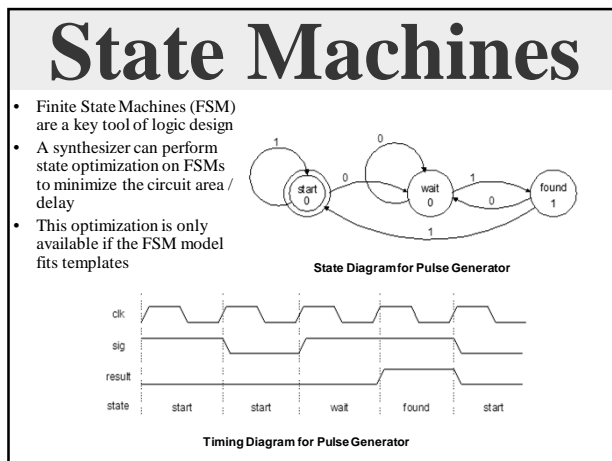




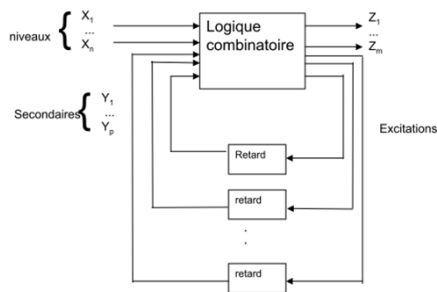
Combinational Logic

```

entity COMB_LOGIC is -- y = c'd+db+cda+c'da'
  generic (P_DELAY : time);
  port (A, B, C, D : In Bit;
        Y : out Bit);
end COMB_LOGIC; -- behavior from the truth table
-- process statement and sequential statements
architecture ARCH2 of COMB_LOGIC is
begin
  process (A, B, C, D)
    variable TEMP : Bit_Vector (3 DOWNTO 0);
  begin
    TEMP := A&B&C&D;
    case TEMP is
      when b'0000' | b'1000' | b'1001 | b'0011' |
           b'0010' | b'0110' | b'1110' | b'1010' =>
        y <= '0';
      when OTHERS => y <= '1';
    end case;
  end process;
end ARCH2;
  
```



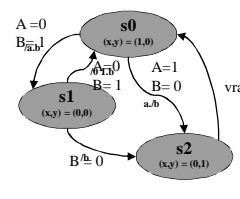
Modèle général (machine asynchrone)



MACHINES d'ETAT (exemple)

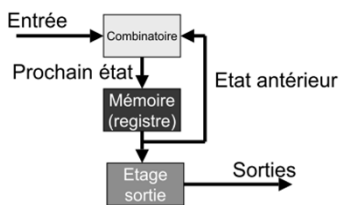
```

ENTITY machine IS
  PORT (a,b,clk : IN BIT;
        x,y : OUT BIT);
END machine;
ARCHITECTURE moore OF machine IS
  TYPE STATE_TYPE IS (s0,s1,s2);
  SIGNAL etat : STATE_TYPE;
BEGIN
  PROCESS
  BEGIN
    WAIT UNTIL clk = '1';
    CASE etat IS
      WHEN s0 => IF a='0' AND b='1' THEN etat <= s1;
                  ELSEIF a='1' AND b='0' THEN etat <= s2;
                  ELSE etat <= s0;
            END IF;
      WHEN s1 => IF a='0' AND b='1' THEN etat <= s0;
                  ELSEIF b='0' THEN etat <= s2;
                  ELSE etat <= s1;
            END IF;
      WHEN s2 => etat <= s0;
      WHEN OTHERS => etat <= s0;
    END CASE;
  END PROCESS;
  x <= '1' WHEN etat = s0 ELSE '0';
  y <= '1' WHEN etat = s2 ELSE '0';
END moore;
    
```

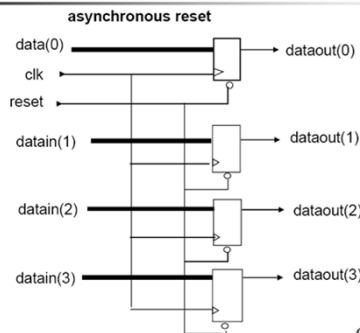


Machine de Moore

- Les sorties dépendent des états présents. Un bloc de logique combinatoire traite les entrées et l'état antérieur des sorties et alimente un bloc de bascules en sortie. Ainsi, les sorties changent de manière synchrone sur un front d'horloge.

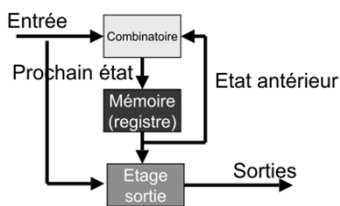


VHDL Register



Machine de Mealy

- A la différence de la machine de Moore, la sortie peut changer lors d'un changement de l'entrée et ce indépendamment de l'horloge.
- Une machine de Mealy est donc asynchrone.
- Une machine de Moore changera seulement sur des transitions du signal d'horloge : son fonctionnement synchrone facilite son utilisation dans tout circuit synchrone, ce qui en fait un avantage sur la machine de Mealy. Il existe cependant une version synchrone de la machine de Mealy.



VHDL Register

```

process (RESET, CLK)
begin
  if RESET = '0' then
    DATAOUT <= "0000";
  elsif CLK'event and CLK = '1' then -- rising edge
    DATAOUT <= DATAIN; -- of clock
  end if;
end process;
    
```

Asynchronous reset does not depend on clock.

LOGI QUE SEQUENTI ELLE / compteurs

compteur: association d'un registre et d'un additionneur

A word of caution. Hardware realization of VHDL objects

```

signal A,B,C,D: bit
...
NO_MEMORY: process (A,B,C)
variable TMP: bit;
begin
    TMP := A and B;
    D <= TMP or C;
end process;
    
```

**Implementation of TMP :
wire**

```

signal A,B,C: bit
...
IS_IT_LATCH: process (A,B,C)
variable TMP: bit;
begin
    C <= TMP and B;
    TMP := A or C;
end process;
    
```

**Implementation of TMP :
latch. Do you real want it?**

LOGI QUE SEQUENTI ELLE / compteurs (sui te)

exemple: compteur 4 bits modulo 10

```

ENTITY compt10 IS
    PORT (CK : IN BIT;
          Q : OUT INTEGER RANGE 0 TO 9);
END compt10;
ARCHITECTURE mod10 OF compt10 IS
    SIGNAL valeur : INTEGER RANGE 0 TO 9;
BEGIN
    PROCESS (CK)
    BEGIN
        IF CK'EVENT AND CK='1' THEN
            IF valeur = /9 THEN valeur <= valeur +1;
            ELSE valeur <= 0;
            END IF;
            ELSE valeur <= valeur;
            END IF;
        END PROCESS;
        Q <= valeur;
    END mod10;
    
```

Séquentialiser si besoin!

- Interdire un feedback zéro délai :
OP <= OP + Y ;
- syntaxiquement correcte, refusée par la plupart des synthétiseurs
- On utilise un process :
Process (OP, Y)
Begin
OP <= OP + Y ;
End process ;

LOGI QUE SEQUENTI ELLE / compteurs (sui te)

pour info: structure physique du compteur 4 bits modulo 10

- synthétisé par le compilateur
- transparent pour le concepteur

VHDL!!!

- C'est la pratique qui fait le designer !